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Pre-insertion Resistors in High Voltage Capacitor Bank Switching

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Abstract: The switching of high-voltage capacitor banks for reactive-power or voltage support can produce significant transients. It is well understood that reactors, pre-insertion resistors, pre-insertion inductors, and synchronous switching can mitigate the transients. Circuit inductance can limit peak currents but resistance damps the oscillations most effectively. Computer simulations of transient inrush for single and back-to-back capacitor banks indicate that pre-insertion resistors can significantly reduce transients. The ability to incorporate another circuit parameter, the pre-insertion resistor, provides opportunities for improved high-voltage capacitor bank design.

This paper presents an overview of capacitor switching options and the results of computer simulations for a "typical" application showing the expected transient currents in single-bank and back-to-back switching, as well as the effect of various reactor and pre-insertion resistor combinations. Results of the simulations are summarized in a table and provide the reader with a simple overview of the results of using pre-insertion resistors in capacitor switching applications.

Capacitor Switching Transients

Capacitor switching transients are created by the effective short circuit provided by a discharged capacitor during energization. This short circuit creates high inrush currents and the subsequent voltage dips on the source. The high inrush currents also stress switching equipment, fuses, and the capacitor units.

When more than one capacitor bank exists on a common bus, the energized capacitor bank provides an extremely low source-impedance for the second switching capacitor bank leading to extremely high transient currents in both banks. In grounded wye banks, more common at higher voltages, these high transient currents can raise ground grid potentials and may damage anything tied to the same ground mat.

The engineer's goal is to control, limit and direct transient currents so that adverse effects on the source system, the capacitors, their switching devices, and associated equipment are limited to acceptable values. Since every switching event has the potential for very high currents, limiting transients also prolongs equipment life. It is hoped that a workable compromise between competing values can be achieved.

When switching transients are limited in grounded banks, the required mitigation for these transients on ground grid design, control cable grounding and routing, and transient over-voltage protection can be simplified.

In addition to the local effects of current and voltage transients during capacitor switching, the remote effects of "voltage magnification" at lower voltage capacitors can be an issue. Several excellent papers on this subject are available.

IEEE Classic Estimates

ANSI/IEEE C37.012-1979 IEEE Application Guide for Capacitance Current Switching for AC High-Voltage Circuit Breakers provides an accepted analytic approach for computation and estimation of the transient currents expected during capacitor switching. However, the complexities of modern equipment including pre-insertion inductors, pre-insertion resistors, full-time inductors, resistance of current limiting fuses are not considered in the basic IEEE methods. As such, it is an excellent guide and first-order approximation to the expected system performance. These equations are:

Single Bank Switching

I max peak = Sqrt(2 * Isc * Ic)

Freq (Hz) = 60 * Sqrt (Isc / Ic)

Back-to-Back Switching

I max peak = 1750 * Sqrt (Vll * Ic1 * Ic2 / (Leq * (Ic1 + Ic2)))

Freq (kHz) = 9.5 * Sqrt (Fs * Vll * (Ic1 + Ic2) / (Leq * Ic1 * Ic2))

Where:
Isc = 3-phase RMS symmetrical short circuit current in Amperes
Ic = RMS current in capacitor bank in Amperes
Vll = Line-to-line voltage in kilo-Volts
Fs = 60 Hz
Leq = Equivalent per-phase inductance between capacitor banks in micro-Henrys

Computer Simulations

To enhance our understanding of capacitor switching transients, modern computer simulations are excellent tools. Power engineers have long used load flow and short circuit modeling software. More sophisticated area control problems make use of system stability software. However, it has been electronic engineers who have taken advantage of simulation programs such as EMTP, PSCAD, PSpice. Such simulation programs allow the engineer to create a model of the electrical system and the control system, integrate the two and observe the effects of changes in system design and control. The Bonneville Power Administration is a supporter and proponent of ATP, a public-domain version of EMTP.

Such models require the user to build the model symbolically, using text or iconic symbols. PSpice is widely available, however, it requires the development of text files and has a relatively steep learning curve. The authors have used CASPOC for similar simulations. CASPOC, by Simulation Research, Netherlands, was designed for the power electronics industry to allow engineers to "build" a power system from resistors, capacitors, switches, etc. The engineer then builds, using logic blocks, the control system. Outputs from the control system can be used to control the electric circuit. CASPOC is iconic rather than text oriented, provides feedback during the simulation, and has a quick learning curve (Valuable for us aging engineers.) CASPOC was used to develop the simulations and analyses presented in this paper.

Base System

To keep the system modeling for this paper within a reasonable set of parameters, a standard or "typical" capacitor bank was selected for modeling. The principles apply to any capacitor bank. Given the ease of use of the software, there is little reason not to create a specific model for each and every capacitor bank being designed or studied.

For this paper we selected the following:

20,000 kVAr 3-phase (100A Ic)
115kVl-l
Grounded wye
25,000A available fault duty (Isc) (500MVA)
One or two identical banks
Assume 20 micro-Henrys bus reactance to each bank

For this system, the IEEE calculations give:

Single Bank Switching

I max peak = Sqrt(2 * Isc * Ic)Sqrt(2 * 25000 * 100) = 2236 A

Freq (kHz) = 60 * Sqrt (Isc / Ic)60 * Sqrt (25000/100) = 949 Hz

Back-to-Back Switching

I max peak = 1750 * Sqrt (Vll * Ic1 * Ic2 / (Leq * (Ic1 + Ic2)))1750 * Sqrt(115 * 100 * 100/((20 + 20) * (100 + 100))) = 20,982 A

Freq (kHz) = 9.5 * Sqrt (Fs * Vll * (Ic1 + Ic2) / (Leq * Ic1 * Ic2))9.5 * Sqrt(60 * 115 * (100 + 100) / ((20 + 20) * 100 * 100)) = 17.6 kHz

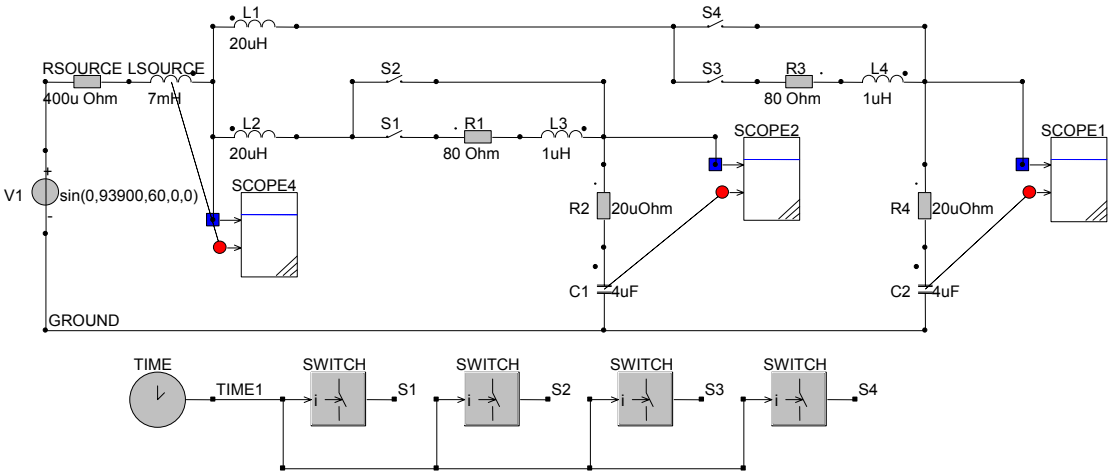
Transient Control Options

Using this "typical" capacitor bank, switching transients were examined using CASPOC. Five system configurations were studied.

No Transient Limiting
Full-time Inductor
Pre-insertion Inductor (Standard & Enhanced)
Zero-Crossing Breaker with 1ms Error
Pre-insertion Resistor (Standard & Enhanced)

For each system, we present typical transient summaries. Our focus is on worse case switching transients since these provide the proof of the design. For each case, we show the transients for switching of the first capacitor bank and the results for the switching of the second bank after the first is already energized. Each bank is independently controlled. Details of the peak current, ringing frequency, and peak voltage are provided for each portion of the switching event.

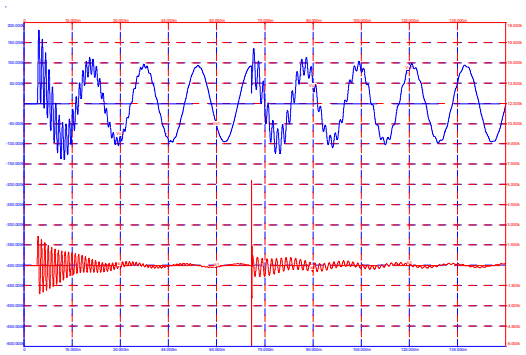
The general diagram for the CASPOC model is shown below. The various models were created by varying component parameters and switch timings.



In this model diagram, inductors, capacitors, and resistors are clearly shown. Oscilloscope blocks are attached to the circuit to monitor voltages and currents. Switches are shown in the circuit and their associated control blocks shown below the schematic. Each switch block has its closing time programmed as a parameter for that block.

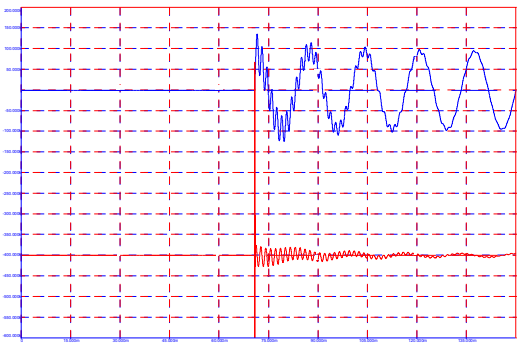
No Transient Limiting Employed

This would be the basic "across the line" switching of the capacitor bank. No attempt is made to reduce or limit the transient inrush into the capacitor bank and through the switching device. Where the source resistance is high or the capacitor bank is small relative to the ratings of the switching devices, this will be the simplest and least cost option. This is the most common switching design for small capacitor banks installed on distribution feeders or in substations with weak sources. Worst-case switching occurs at the peak of the cycle in the simulation.



Voltage and Current for Energization of First Bank

No Transient Limiting	Peak Current	Frequency	Peak Voltage
Bank 1 Energization	2172A	943Hz	182kV (1.93pu)
Bank 2 Energization	14038A	16400Hz	137kV (1.46pu)
Bank 2 Ringing	830A	681Hz	

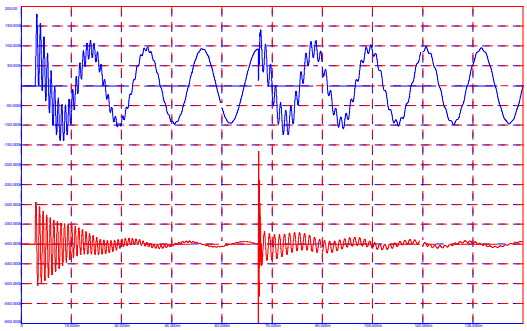


Voltage and Current for Energization of 2nd Bank

No Transient Limiting	Peak Current	Frequency	Peak Voltage
Bank 2 Energization	14149A	16400Hz	137kV (1.46pu)
Bank 2 Ringing	830A	674Hz	

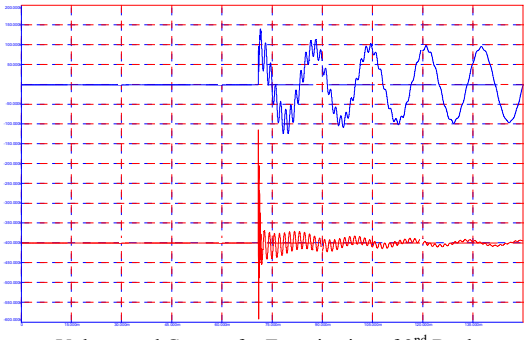
Full-time Inductor

A common design at lower voltages and chosen for its simplicity is the use of a full-time inductor in the switched circuit. The inductor is chosen to limit inrush below the damage levels for the switching device, the capacitor units and any associated equipment. However, there is a tradeoff since the inductor is continuously energized and produces losses (heat) from its resistance. To limit peak currents to 6000A, the inductor is 200 micro-Henry and about 20milli-Ohm.



Voltages and Current for Energization of 1st Bank

Full Time Inductor	Peak Current	Frequency	Peak Voltage
Bank 1 Energization	2155A	925Hz	182kV (1.94pu)
Bank 2 Energization	5459A	5329Hz	143kV (1.52pu)
Bank 2 Ringing	806A	670Hz	

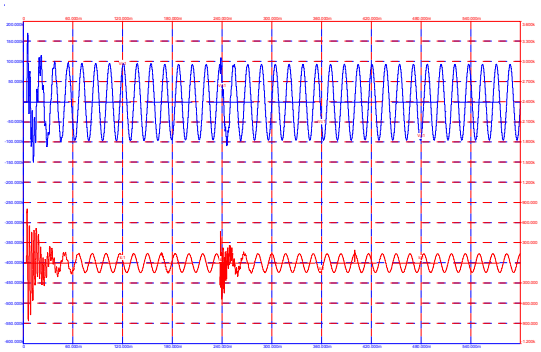


Voltage and Current for Energization of 2nd Bank

Full Time Inductor	Peak Current	Frequency	Peak Voltage
Bank 2 Energization	5690A	5310Hz	142kV (1.51pu)
Bank 2 Ringing	846A	662Hz	

Pre-insertion Inductor

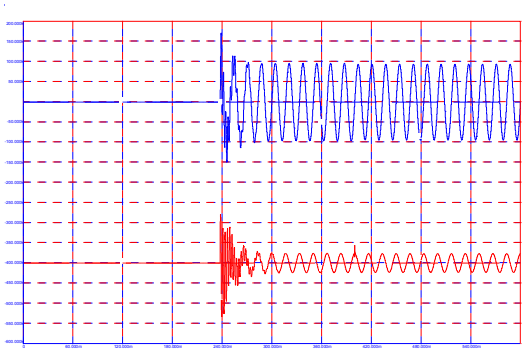
In this system, a two-stage switching device is used to momentarily introduce an inductance into the circuit before shorting out the inductance by the main switching device contacts. An open-air contact makes the initial circuit through the pre-insertion inductor. The pre-insertion inductor is bypassed and disconnected as the switch blade rotates into its contacts. The inductor (reactor) is chosen to reduce the initial transient upon energizing the voltage-less capacitor bank and to balance the inrush events between this initial inrush and the inrush from bypassing the inductor. The inductor is typically 40mH, 5.5 Ohm, and is in the circuit for 7-12 cycles (117-200mS). Worst case transients occur when the initial switch closing occurs at a voltage peak and the bypassing of the inserted device occurs at a current peak. Simulations were performed using this timing. Manufacturer's also offer an "enhanced" option of 40mH and 81 Ohms. Results for both options are presented below.



Voltage and Current for Energization of 1st Bank at 5.5 Ohms

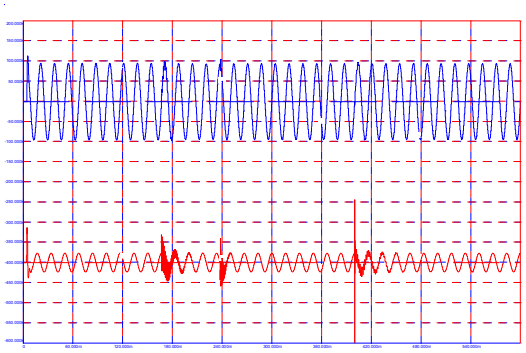
Pre-Insertion Inductor	Peak Current	Frequency	Peak Voltage
Bank 1 Energization	812A	365Hz	172kV (1.83pu)
Bank 2 Energization	540A	1036Hz	111kV (1.18pu)

During the time that the pre-insertion inductor is in the circuit, heat is being generated in this device. The magnitude of the I^2t times the resistive value of the pre-insertion inductor gives an indication of the transient heat that must be handled and dissipated. The I^2t on the 5.5 ohm, 40mH, pre-insertion inductor is $4500A^2sec$.



Voltage and Current for Energization of 2nd Bank at 5.5 Ohms

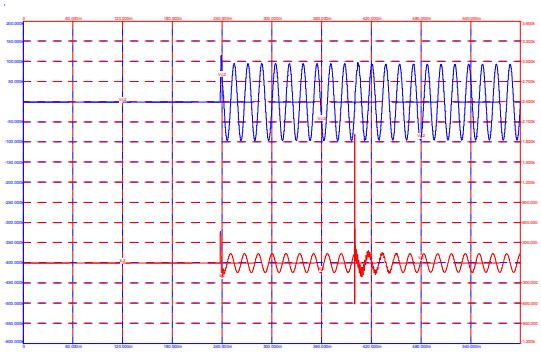
Pre-Insertion Inductor	Peak Current	Frequency	Peak Voltage
Bank 2 Energization	810A	361Hz	149kV (1.58pu)
Bank 2 Transient	268A	16,340Hz	97kV (1.03pu)
Bank 2 Ringing	147A	657Hz	94.5kV (1.00pu)



Voltage and Current for Energization of 1st Bank at 81 Ohms

Pre-Insertion Inductor	Peak Current	Frequency	Peak Voltage
Bank 1 Energization	526A	351Hz	114kV (1.21pu)
Bank 1 Bypass Closes	419A	948Hz	101kV (1.08pu)
Bank 2 Energization	365A	1030Hz	107kV (1.14pu)
Bank 2 Transient	1620A	16130Hz	98.5kV (1.05pu)
Bank 2 Ringing	240A	672Hz	98.5kV (1.05pu)

The I^2t on the 81 ohm, 40mH, pre-insertion inductor is 1900A^2sec.

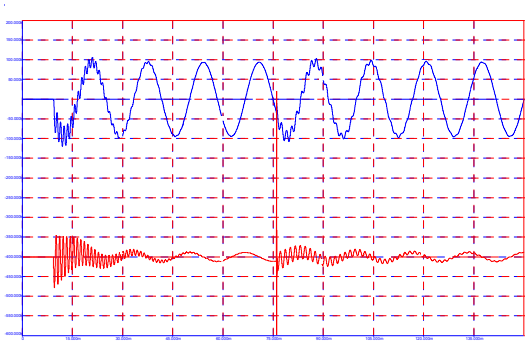


Voltage and Current for Energization of 2nd Bank at 81 Ohms

Pre-Insertion Inductor	Peak Current	Frequency	Peak Voltage
Bank 2 Energization	479A	1034Hz	116kV (1.24pu)
Bank 2 Transient	1920A	16529Hz	98.5kV (1.05pu)
Bank 2 Ringing	210A	670Hz	

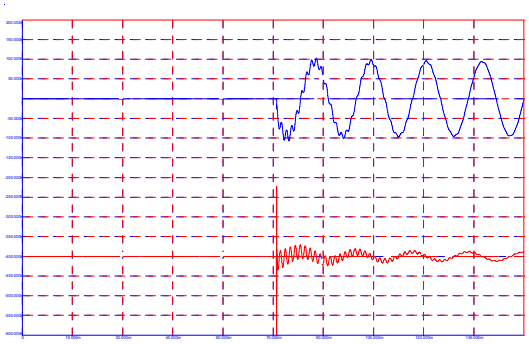
Zero-crossing Breaker

Modern control systems attempt to exploit the ability to precisely and repetitively control with precision the instant at which the switching contacts come together. Under ideal circumstances, if the poles close at the point of zero-voltage, there will be no current transients created. This requires precise timing and control of the three individual poles. Any drift in the control must be compensated for or else the system reverts toward the first case above. Simulations for 1mS closing error are performed. This error is chosen for simulation purposes only and is not meant to be indicative of the actual zero-crossing performance of any specific breaker or installation.



Voltage and Current for Energization of 1st Bank

Zero-crossing Breaker	Peak Current	Frequency	Peak Voltage
Bank 1 Energization	942A	944Hz	120kV (1.28pu)
Bank 2 Transient	5021A	16807Hz	108kV (1.15pu)
Bank 2 Ringing	419A	672Hz	

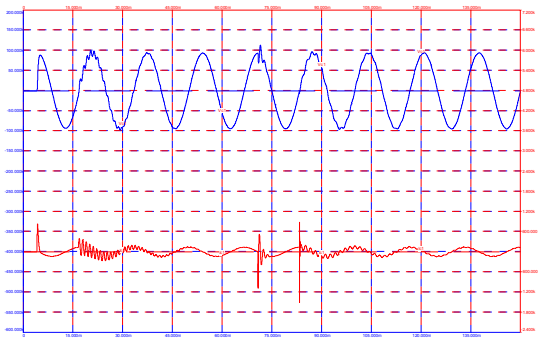


Voltage and Current for Energization of 2nd Bank

Zero-crossing Breaker	Peak Current	Frequency	Peak Voltage
Bank 2 Energization	5228A	16,667Hz	108kV (1.15pu)
Bank 2 Ringing	420A	670Hz	

Pre-insertion Resistor

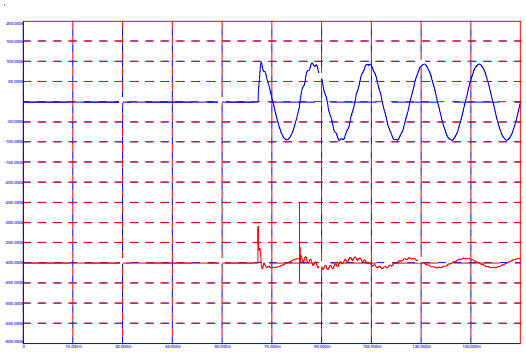
Similar to the pre-insertion inductor, the pre-insertion resistor provides inrush limiting by the momentary insertion of a resistive device into the circuit before full energization of the capacitor bank. The insertion of the resistor is a two step process. The initial circuit is made through the pre-insertion resistor in an SF6 environment. The resistor is then shunted as the main contacts close. For similar levels of transient suppression, the pre-insertion resistor can be physically smaller than the equivalent pre-insertion inductor. Various values of pre-insertion resistor are available. A typical manufacturer-recommended pre-insertion resistor for this application is 80 Ohm and is in the circuit for 5-15mS. An enhanced 150 Ohm pre-insertion resistor is also available. Worst case transients occur when the initial switch closing occurs at a voltage peak and the bypassing of the inserted device occurs at a current peak. Simulations were performed using this timing. Results for both options are included below.



Voltage and Current for Energization of 1st Bank at 80 Ohms

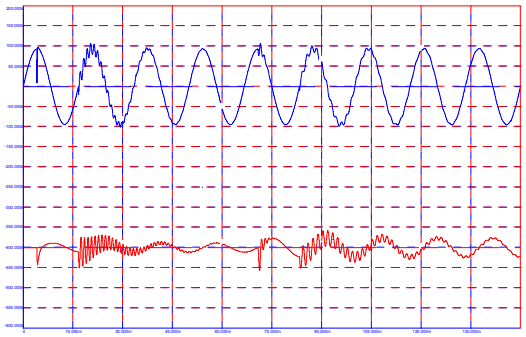
Pre-insertion Resistor	Peak Current	Frequency	Peak Voltage
Bank 1 Energization	835A	NA	97kV (1.03pu)
Bank 1 Transient	404A	948Hz	101kV (1.07pu)
Bank 2 Energization	1100A	809Hz	114kV (1.21pu)
Bank 2 Transient	1520A	16,400Hz	
Bank 2 Ringing	235A	670Hz	97kV (1.03pu)

In addition, the I^2t for the 80 ohm pre-insertion resistor is $330A^2s$.



Voltage and Current for Energization of 2nd Bank at 80 Ohms

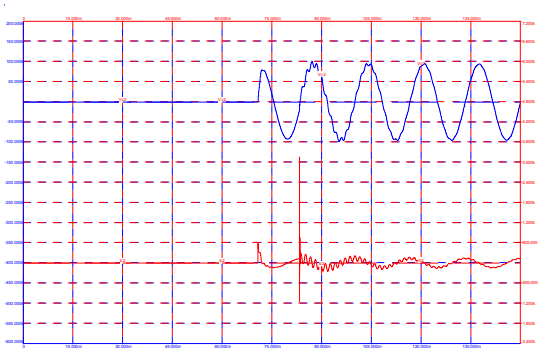
Pre-insertion Resistor	Peak Current	Frequency	Peak Voltage
Bank 2 Energization	1100A	892Hz	98.7kV (1.05pu)
Bank 2 Transient	1820A	16,529Hz	
Bank 2 Ringing	235A	672Hz	98.3kV (1.05pu)



Voltage and Current for Energization of 1st Bank at 150 Ohms

Pre-insertion Resistor	Peak Current	Frequency	Peak Voltage
Bank 1 Energization	529A	NA	
Bank 1 Ringing	614A	947Hz	107kV (1.14pu)
Bank 2 Energization	600A	903Hz	108kV (1.15pu)
Bank 2 Transient	2870A	16,400Hz	
Bank 2 Ringing	312A	670Hz	102kV (1.09pu)

In addition, the I^2t for the 150 ohm pre-insertion resistor is $215A^2s$.



Voltage and Current for Energization of 2nd Bank at 150 Ohms

Pre-insertion Resistor	Peak Current	Frequency	Peak Voltage
Bank 2 Energization	600A	860Hz	
Bank 2 Transient	3170A	16,529Hz	
Bank 2 Ringing	312A	671Hz	102kV (1.09pu)

Pre-Insertion Resistor Actual Field Results

In order to verify the basic system modeling approach, actual field test results for operation of a pre-insertion resistor switch were obtained courtesy of a U. S. investor owned utility. The actual transient records (oscillograph results attached) were compared to a simulation model (results attached). The results agree favorably in that peak transient currents are of the same magnitude and ringing frequencies are similar.

The actual results show lower transients magnitudes and less ringing upon pre-insertion resistor bypass. These can be explained by the lack of detailed component data used in to build the model. Higher actual component resistances than those used in the model will lead to reduced transients. Substation load will cause some additional damping of transients. Also, current transformer hysteresis may add to current measurement error at the 4500Hz ringing frequencies.

Generally, the model provides good agreement with the recorded data.

Summary

The following table summarizes the most significant data in the above results:

Simulation Case	Single-bank Peak Current	Single-bank Frequency	Back-to-back Peak Current	Back-to-back Frequency	Peak Voltage
IEEE Calculations	2236A	949Hz	20,982A	17,600Hz	
No damping	2172A	943Hz	14,038A	16,400Hz	1.93pu
Full time inductor	2155A	925Hz	5459A	5329Hz	1.94pu
Standard pre-insertion inductor	812A	365Hz	810A	16.340Hz	1.83pu
Enhanced pre-insertion inductor	526A	351Hz	1920A	16,529Hz	1.23pu
Zero crossing with 1mS error	942A	944Hz	5021A	16,807Hz	1.28pu
81 Ohm pre-insertion resistor	835A	948Hz	1820A	16,529Hz	1.22pu
150 Ohm pre-insertion resistor	614A	947Hz	3170A	16,529Hz	1.14pu

Conclusions

The conclusions we would like to take from this work are:

1) IEEE guidelines for calculation of capacitor switching transients are good first estimates for transients with no limiting applied.

2) The engineer has several options for limiting transient events during capacitor switching.

   a. The full time inductor, pre-insertion inductor, zero-crossing breaker, and pre-insertion resistor successfully mitigate current transients.

   b. Devices with resistance provide the added advantage of reducing voltage transients. Proper selection of the resistance value can significantly reduce voltage transients.

   c. Zero-crossing switching shows good transient mitigation but transients will increase if timing calibration drifts.

   d. The transients from connection of the pre-insertion device and those from bypassing the pre-insertion device need to be considered.

3) Transient modeling, by computer simulation software, has produced results that are consistent with actual field testing.

4) Modern computer simulation software can provide the engineer with a better understanding of the transient events during capacitor switching.

5) Because every capacitor bank and electric system is different, the designer must analyze the situation and tailor the switching device accordingly. Manufacturer's offer variations to aid the designer in this customization.

References

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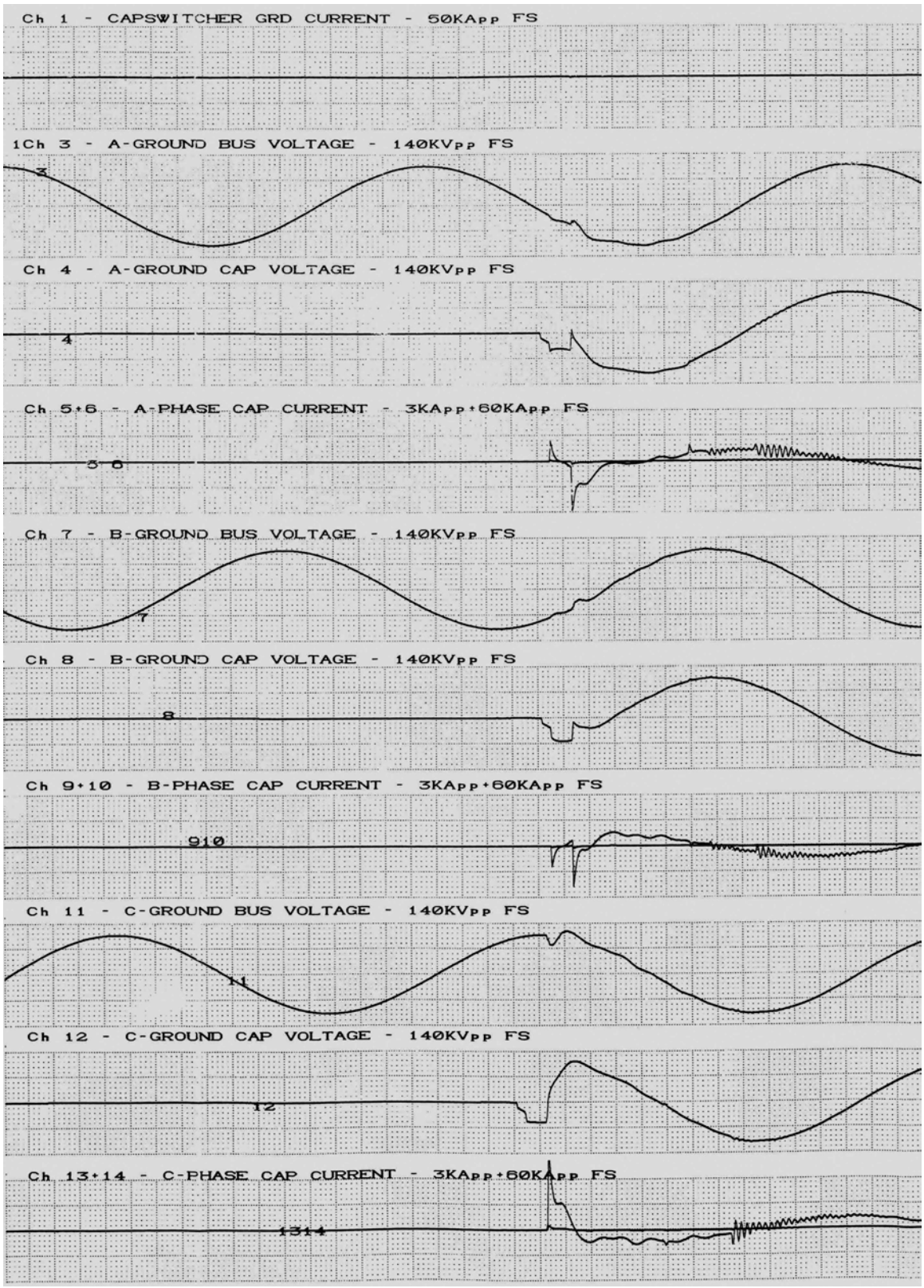
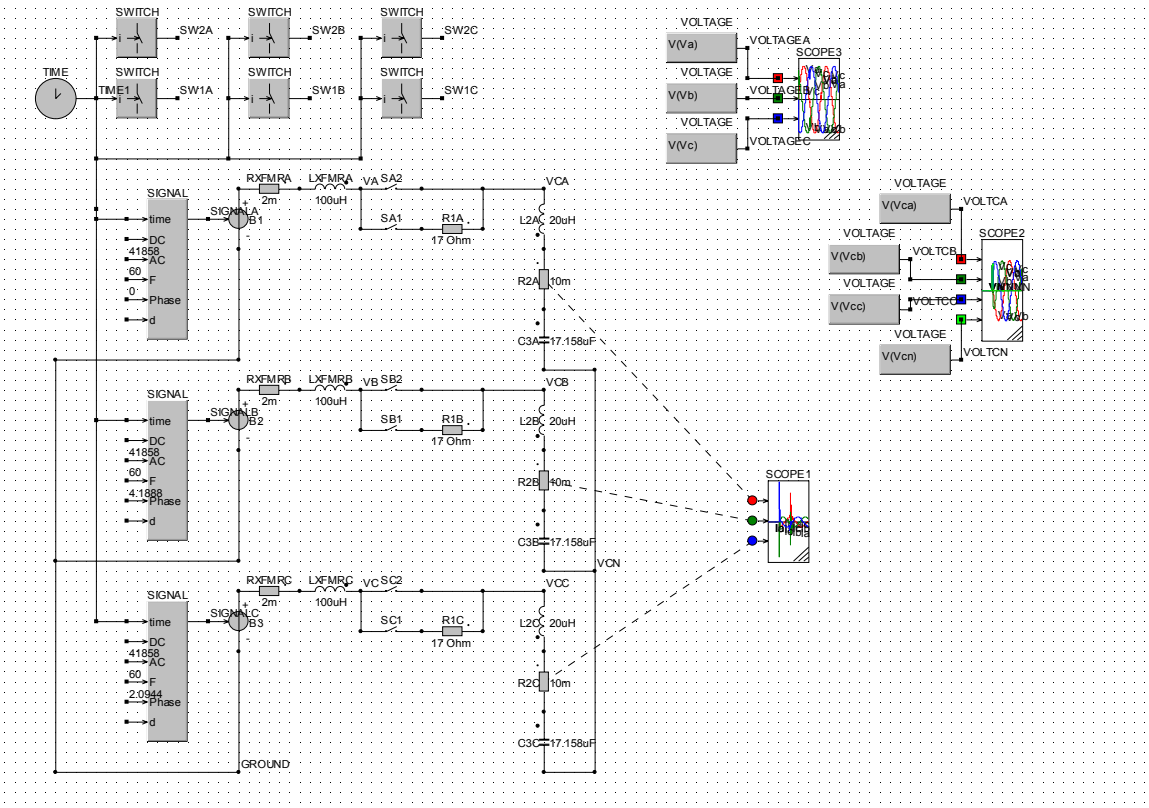
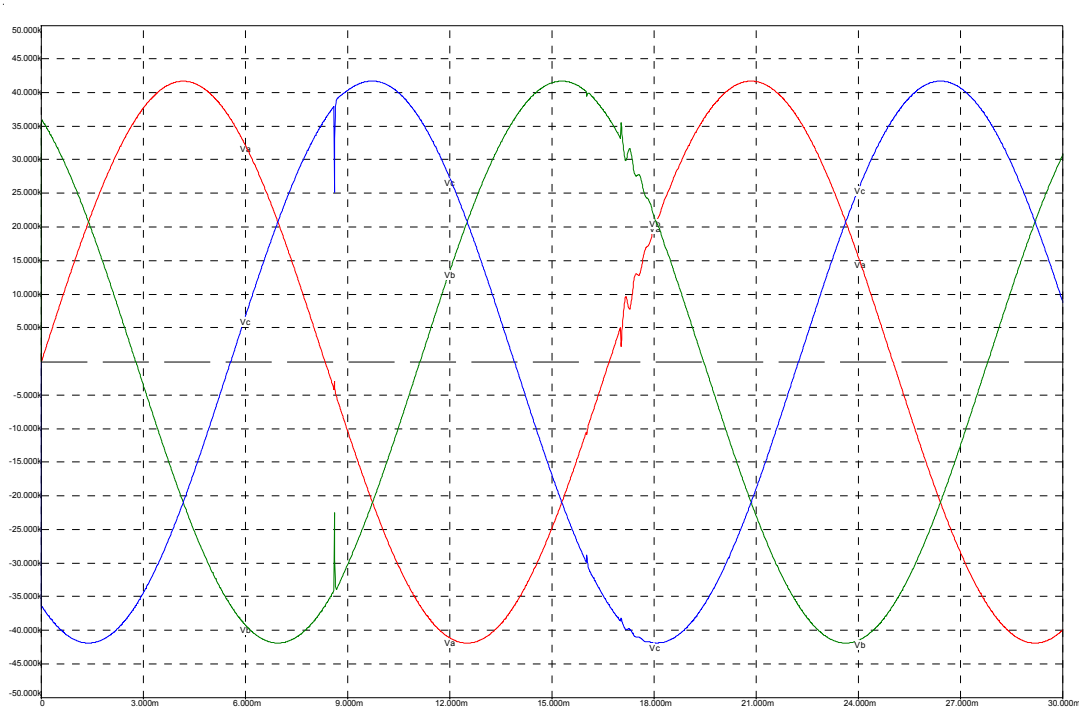


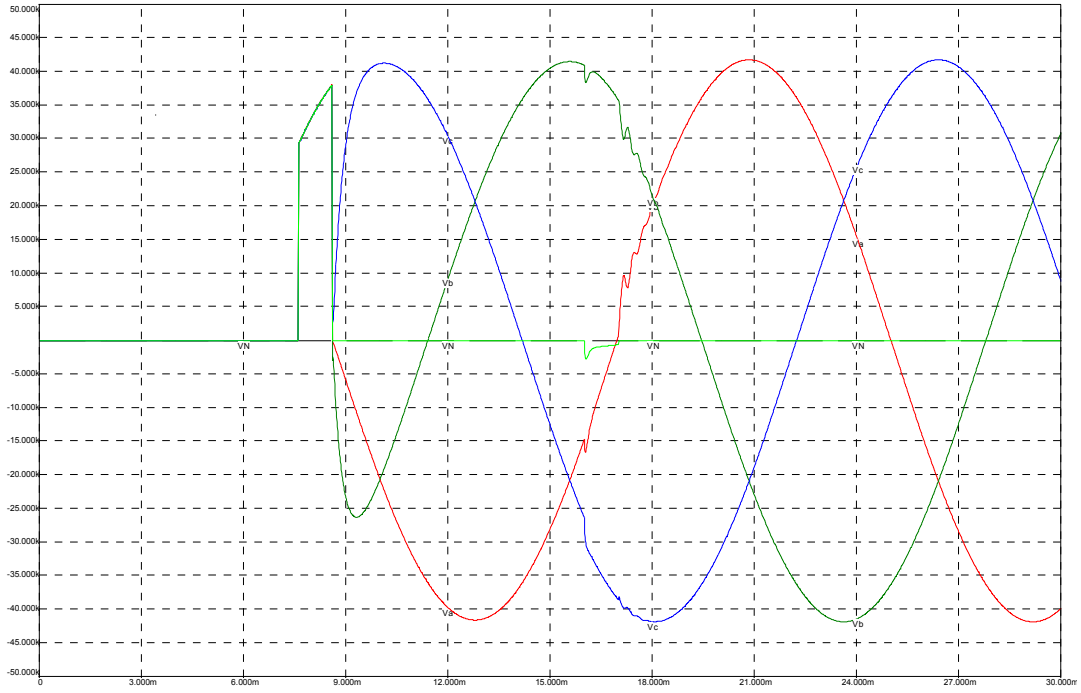
Oscillograph from Installation at U. S. Utility
For ***CapSwitcher***® Capacitor Switch with Pre-Insertion Resistors



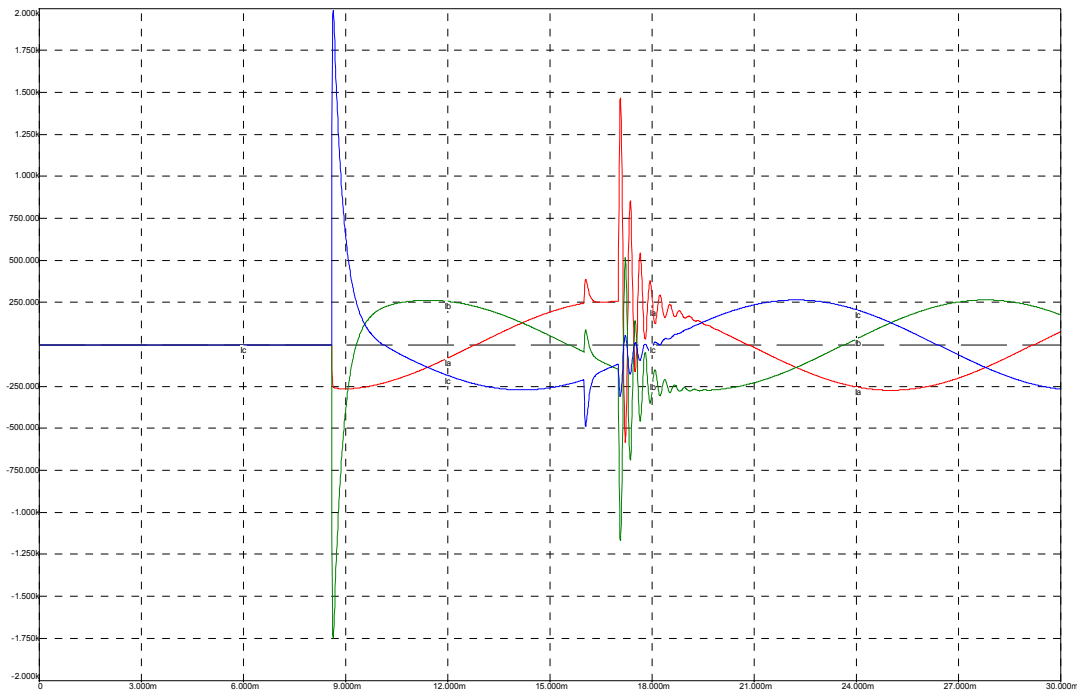
Simulation Model for Ungrounded Wye Capacitor Bank
72.5kV, 27.5MVAr installed at U. S. Utility



Substation Bus Voltages from Simulation for U. S. Utility
Installation of Capacitor Switch using Pre-Insertion Resistor



Capacitor Voltages from Simulation for U. S. Utility
Installation of Capacitor Switch using Pre-Insertion Resistor



Capacitor Currents from Simulation for U. S. Utility
Installation of Capacitor Switch using Pre-Insertion Resistor

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