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**USING THERMAL LIMIT CURVES TO DEFINE THERMAL  
MODELS OF INDUCTION MOTORS**

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# USING THERMAL LIMIT CURVES TO DEFINE THERMAL MODELS OF INDUCTION MOTORS

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## INTRODUCTION

IEEE Standard C37.96-2000, *Guide for AC Motor Protection* [1], recommends the use of overcurrent relays for overload and locked rotor protection. In these applications, setting the overcurrent inverse time-current characteristic to coordinate with the motor thermal limit curves provides protection. Because of the familiar use of overcurrent protection, little attention is paid to the nature of motor thermal limit curves and their relation to winding temperature in an induction motor. Yet, the limit curves are the characteristics of thermal models that enable microprocessor relays to continuously calculate and monitor motor temperature in real time. This paper uses induction motor thermal limit curves to define the thermal model and correlate motor current with temperature. The paper employs MATLAB simulations to compare the dynamics of the thermal model with that of an overcurrent implementation during cyclic overloads.

## THERMAL LIMIT CURVES

The thermal limit curves for a 400-horsepower, 3600-rpm, 440-volt induction motor are shown in Figure 1. By definition, the curves give the time for current exceeding the service factor to raise the initial load temperature to an overload temperature that requires the motor to be disconnected. The service factor (SF) in this case is 1.15. The manufacturer has specified the initial temperature for a set of “hot” and “cold” overload and locked rotor curves.

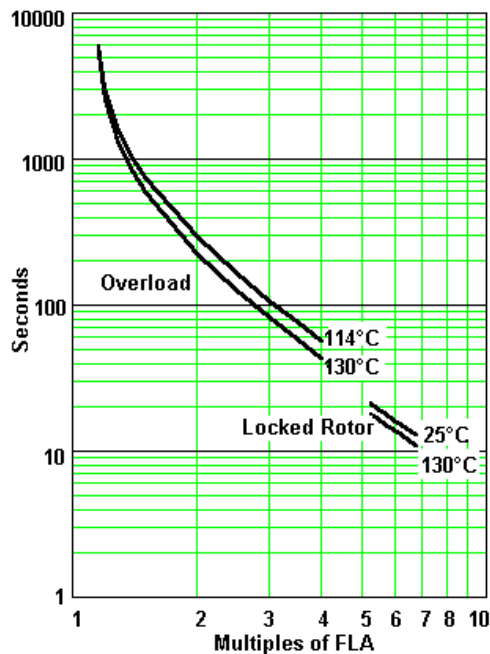


Figure 1 Hot and Cold Limit Curves for a 400-HP Motor, SF = 1.15

The initial motor temperature values are marked on each curve. The curves are presented in accordance with guidelines in IEEE Standard 620-1996, *Guide for the Presentation of Thermal Limit Curves for Squirrel Cage Induction Motors* [2]. Otherwise, the guide gives no information as to how the curves are constructed. However, we can conclude that each curve is a plot of a specific limiting temperature.

## DEFINING THE THERMAL MODELS

The basic equations found in thermal protection modeling are derived in the Annex. Equation A21 gives us the form for overload thermal limit curves plotted in Figure 1. The equations for the overload curves of Figure 1 have the form:

$$\begin{aligned} t_{H-CURVE} &= T_{th} \ln \left( \frac{I^2 - I_H^2}{I^2 - I_{SF}^2} \right) \\ t_{C-CURVE} &= T_{th} \ln \left( \frac{I^2 - I_C^2}{I^2 - I_{SF}^2} \right) \end{aligned} \quad (1)$$

where:  $T_{th}$  is the thermal time constant.

$I$  is the motor current in per unit of full load.

$I_{SF}$  is the current at the service factor.

$I_H$  is the current that raised the temperature to 130°C.

$I_C$  is the current that raised the temperature to 114°C.

If the curves obey a first order thermal process, we will be able to choose  $T_{th}$ ,  $I_H$ , and  $I_C$  so that equations fit the curves. Also,  $I_H$  and  $I_C$  must be in the ratio of the initial temperatures above ambient. A unique solution is obtained under the necessary constraint that:

$$\frac{I_H^2}{I_C^2} = \left( \frac{130 - 25}{114 - 25} \right) = 1.179 \quad (2)$$

$$I_C^2 = \frac{I_H^2}{1.179} \quad (3)$$

The fit is obtained by satisfying simultaneous Equations 4 where the equations have been solved for the service factor  $SF^2$ . The curve fitting procedure is as follows:

1. Choose a current and read the corresponding time points from the hot (130°C) and the cold (114°C) overload curves in Figure 1. Enter the current and time values in Equations 4. For example, at 2-per-unit current, the hot and cold times are  $t_{H-CURVE} = 223$  seconds and  $t_{C-CURVE} = 279$  seconds, respectively.
2. Choose the  $T_{th}$  and  $I_H$  so that Equations 4 equal the service factor.
3. Cut and try to refine the values. For example, when  $T_{th} = 1370$  seconds and  $I_H = 0.92$  per unit, both equations equal  $1.15^2 = 1.322$ .
4. Check the result at other points on the curve. Curve values are shown in Table 1.

$$\begin{aligned}
I^2 \left( 1 - e^{-\frac{t_{H-CURVE}}{T_{th}}} \right) + I_H^2 \cdot e^{-\frac{t_{H-CURVE}}{T_{th}}} &= 1.15^2 \\
I^2 \left( 1 - e^{-\frac{t_{C-CURVE}}{T_{th}}} \right) + \frac{I_H^2}{1.179} e^{-\frac{t_{C-CURVE}}{T_{th}}} &= 1.15^2
\end{aligned} \tag{4}$$

**Table 1** Thermal Limit Curve Checkpoints

<b>I (Per Unit)</b>	<b>t<sub>H-CURVE</sub> (Seconds)</b>	<b>t<sub>C-CURVE</sub> (Seconds)</b>	<b>T<sub>th</sub> (Seconds)</b>	<b>I<sub>H</sub><sup>2</sup> (Per Unit)</b>	<b>I<sub>C</sub><sup>2</sup> (Per Unit)</b>	<b>I<sub>SF</sub> (Per Unit)</b>
2.0	223	279	1370	0.846	0.717	1.15
2.5	126	158	1370	0.846	0.717	1.15
3.0	82	104	1370	0.846	0.717	1.15

When the unique values for  $T_{th}$ ,  $I_H$ , and  $I_C$  are used, Equations 5 can be used to calculate any curve point with precision:

$$\begin{aligned}
t_H &= 1370 \cdot \ln \left( \frac{I^2 - 0.846}{I^2 - (1.15)^2} \right) \\
t_C &= 1370 \cdot \ln \left( \frac{I^2 - 0.717}{I^2 - (1.15)^2} \right)
\end{aligned} \tag{5}$$

Equations 5 are the solutions of a first order differential equation that we will use to implement the thermal model:

$$I^2 R_T = R_T C_T \cdot \frac{d\theta}{dt} + \theta \tag{6}$$

where:  $R_T C_T$  is the thermal time constant = 1370 seconds

$\theta$  = temperature at  $t = 0$

$R_T = 1$

The time-discrete form of Equation 6 can be written as:

$$I^2 R_T = R_T C_T \cdot \frac{\theta_n - \theta_{n-1}}{\Delta t} + \theta_{n-1}$$

Solving for  $\theta_n$  gives the time-discrete form of the differential equation:

$$\theta_n = \frac{I^2}{C_T} \Delta t + \left( 1 - \frac{\Delta t}{R_T C_T} \right) \cdot \theta_{n-1} \tag{7}$$

Equation 7 is the algorithm that enables a microprocessor relay to continuously calculate motor temperature. The temperature is then compared to predetermined trip and alarm thresholds to provide thermal protection under any condition of operation.

### **Thermal Versus Overcurrent Model**

It is instructive to compare response of the thermal model to an overcurrent model of a thermal limit curve. The overcurrent model is implemented by integrating the reciprocal of the hot thermal limit curve as specified in Equation 3 of the IEEE Standard C37.112-1996 [3]. The incremental equations for this process are:

$$t_H = 1370 \cdot \ln \left( \frac{I^2 - .846}{I^2 - (1.15)^2} \right) \quad (8)$$

For  $I > 1.15$ :

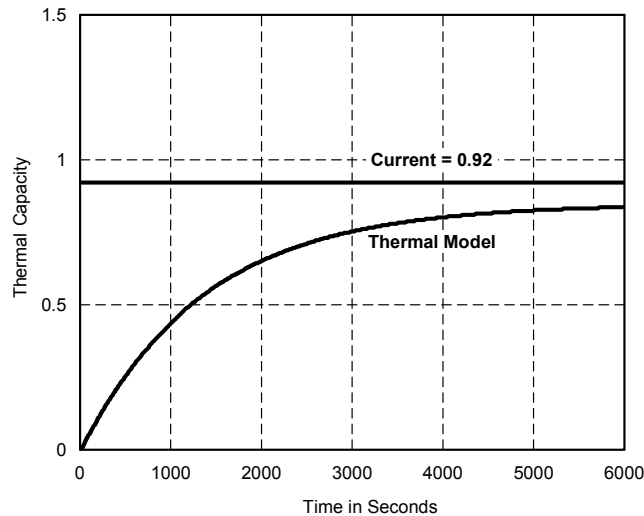
$$\theta_n = \theta_{n-1} + \frac{\Delta t}{t_H} \quad (9)$$

For  $I \leq 1.15$ :

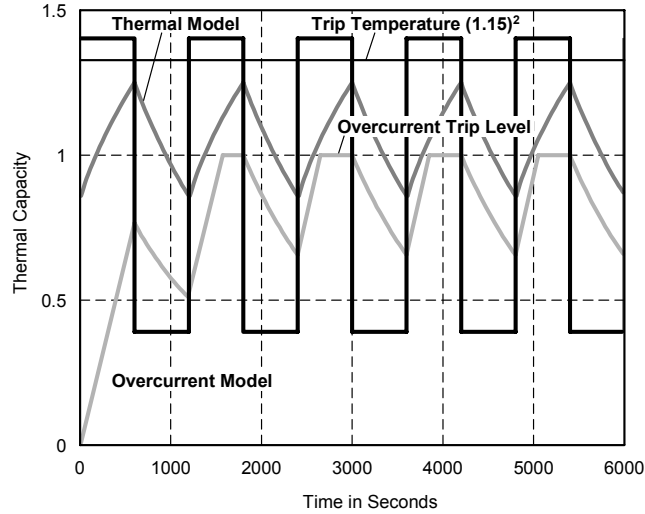
$$\theta_n = \left( 1 - \frac{\Delta t}{1370} \right) \cdot \theta_{n-1} \quad (10)$$

Equation 9 is used to calculate the response of the overcurrent relay above the pickup current. Equation 8 is the time-current characteristic.  $\theta_n$  and  $\theta_{n-1}$  are consecutive samples displaced by one time increment. Below pickup, the overcurrent relay resets exponentially using the thermal time constant to emulate the cooling of the motor.

Figure 2 shows the response of both models to a current below pickup. Whereas the overcurrent model has no response, the thermal model calculates the temperature that rises exponentially toward the steady state temperature  $\theta = 0.846$ . In Figure 3, both models are subjected to a cyclic overload with the motor initially at 0.846 per unit of thermal capacity.



**Figure 2** Response of Models to a Current Less Than the Service Factor ( $I = 0.92$ )



**Figure 3** Response of the Models to a Cyclic Overload Current

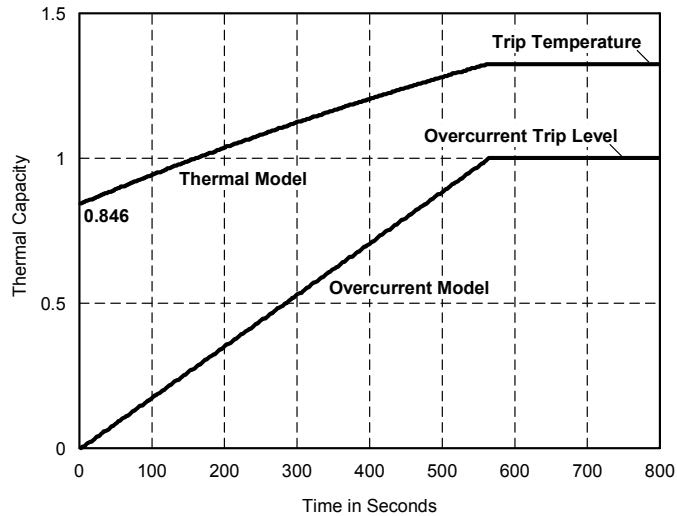
The cyclic current in Figure 2 alternates between 1.4- and 0.4-per-unit current every 10 minutes. Note that the average of the currents squared and the rms current is:

$$I_{2\text{aver}}^2 = \frac{I_{\text{high}}^2 + I_{\text{low}}^2}{2} = \frac{1.4^2 + .4^2}{2} = 1.06 \quad (11)$$

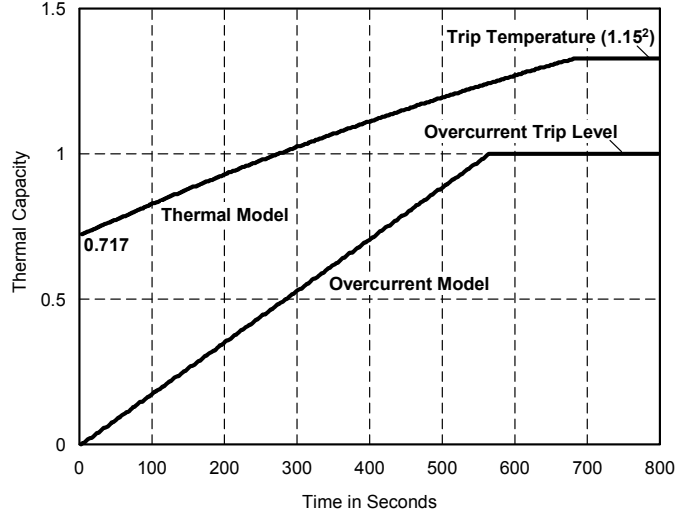
$$I_{\text{RMS}} = \sqrt{1.06} = 1.03$$

The cyclic current is not an overload that raises the temperature to the trip value. Figure 3 shows the cyclic temperature response of the thermal model reaches a 1.06 average, or 80 percent of the trip value.

The overcurrent relay model does not measure temperature and trips because it cannot account for thermal history.



**Figure 4** Thermal and Overcurrent Model Response With Initial Temperature = 0.846



**Figure 5** Thermal and Overcurrent Model Response With Initial Temperature = 0.717

Figure 4 shows that the overcurrent and the thermal model produce the same trip time when the motor is initially at the temperature specified for the hot thermal limit curve. Figure 5 shows the trip times when the motor is initially at the temperature specified for the cold thermal limit curve. The thermal model trips at the limiting temperature while the overcurrent trips at the hot curve time, independent of the initial temperature of the motor.

### **Correlation of Current With Temperature**

The initial temperatures for the “hot” and “cold” overload thermal limit curves are given as 130°C and 114°C, respectively. Also, Equation 5 give the squares of the initial currents as 0.846 and 0.717 per unit, which can be read as per-unit watts in the thermal model. Since temperature is proportional to current squared, we can write:

$$\frac{\theta}{I^2} = \frac{(130 - 25) - (114 - 25)}{0.846 - 0.717} \quad (12)$$

Where  $\theta$  is the motor temperature rise above ambient and  $I$  is current in per unit, we can then write:

$$\theta = \frac{(130 - 25) - (114 - 25)}{0.846 - 0.717} I^2 + 25 = 124.031 \cdot I^2 + 25 \quad (13)$$

Where is  $\theta$  the steady state temperature in degrees centigrade for a given per-unit current, tabulated values are as shown in Table 2.

**Table 2** Correlation of Current to Temperature

<b>Motor Current (Per Unit)</b>	<b>Temperature</b>
1.15 (Service Factor)	189°C
0.92	130°C
0.85	114°C
0.0	25°C

## **CONCLUSIONS**

1. In this paper, we introduced the basic principles and mathematical equations found in thermal protection modeling.
2. We show that the first order thermal equation fits the thermal limit curves of a 400-horsepower, 3600-rpm, 440-volt motor. We then used the hot and cold thermal model equations and the initial temperature data to derive the time constant of the thermal model.
3. We showed the hot thermal limit equation implemented as an overcurrent model. Using MATLAB simulations, we compared the dynamic responses of the overcurrent and thermal models.
4. The simulations show that the thermal model determines the true temperature caused by cyclic overload current, while the overcurrent model trips for cyclic currents that do not overheat the motor.
5. Finally, we use the initial current for the hot and cold equations and the initial temperature data to correlate the motor current with temperature in degrees centigrade.

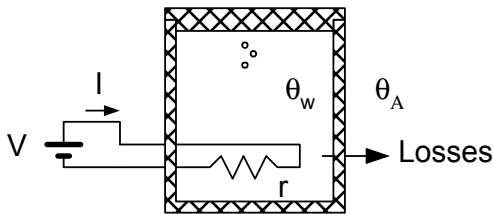


## ANNEX: BASICS OF THERMAL MODELING

This annex introduces the basic principles and mathematical equations found in thermal protection modeling by resolving a simple first-order thermal process: the heating by a resistor of a vessel containing 1 liter of water.

### A Simple Thermal System

Let us assume a vessel contains 1 liter (or 1 kilogram) of water. The water is heated by a source of energy  $V$ .  $\theta_A$  is the ambient temperature surrounding the vessel and  $\theta_w$  is the water temperature. We do not want the water to go beyond the boiling point, or 100 degrees centigrade. This temperature is defined as the maximum or hot-spot temperature  $\theta_{max}$  and above this point, the voltage source will be disconnected by some protective device.



**Figure A1** A Simple Thermal System

Let us define  $\theta$  as the temperature of the water above ambient or:

$$\theta = \theta_w - \theta_A \quad (A1)$$

The rate of increase of the water temperature is provided by the equation expressing the thermal equilibrium.

$$\text{Power Supplied to the Water} - \text{Losses} = C_s m \frac{d\theta_w}{dt} = C_s m \frac{d\theta}{dt} \quad (A2)$$

In this equation,  $C_s$  is the specific heat of the water. It corresponds to the amount of energy that has to be supplied to 1 kilogram of water to raise its temperature by 1 degree centigrade. Its value is 4.19E3 joules-kg/°C.  $m$ , in kilograms, is the mass of the water. The losses or the quantity of heat transferred by the vessel to the surrounding environment can be expressed as:

$$\text{Losses} = \frac{\theta_w - \theta}{R} \quad (A3)$$

Equation A2 can be otherwise expressed as:

$$I^2 r - \frac{\theta}{R} = C_s m \frac{d\theta}{dt} \quad (A4)$$

or:

$$I^2 r \cdot R = C_s m \cdot R \frac{d\theta}{dt} + \theta \quad (A5)$$

The mass  $m$  multiplied by the specific heat  $C_s$  is known as  $C$ , the thermal capacity of the system with units of joules/°C. It represents the amount of energy in joules required to raise the system temperature by 1 degree centigrade.

The product of the thermal resistivity  $R$  and the thermal capacity  $C$  has unit of seconds and represents the thermal time constant  $T_{th}$  of the system:

$$T_{th} = R \cdot m \cdot C_s \quad (A6)$$

The solution in the time domain for the temperature as a function of time and current is:

$$\theta = I^2 r \cdot R \left( 1 - e^{-\frac{t}{R \cdot m \cdot C_s}} \right) \quad (A7)$$

Remembering that  $\theta$  is the temperature above ambient, we obtain for the expression of the water temperature:

$$\theta_w(t) = I^2 r \cdot R \left( 1 - e^{-\frac{t}{T_{th}}} \right) + \theta_A \quad (A8)$$

Let us assume the system has some rated operating current  $I_{op}$  (which is otherwise called the load current in some applications). Then the water operating temperature is given by:

$$\theta_{op} = I_{op}^2 r \cdot R + \theta_A \quad (A9)$$

Whatever the current supplied to the vessel, there will always be an increase in the water temperature. The final water temperature for a constant current  $I$  is given by:

$$\theta_w(0 \rightarrow \infty) = I^2 r \cdot R + \theta_A \quad (A10)$$

Equation A5 is a first order differential equation and has an electrical equivalent: an RC circuit supplied by a current source. The power supplied to the water in the thermal process is equivalent to the current source supplying the RC circuit. The temperature in the thermal process is equivalent to the voltage across the capacitor in the RC circuit. The equivalence between the two systems is shown in Figure A2.

When supplied by a current step function, the temperature or voltage time response is shown in the same figure.

### **The Equivalent Time-Current Curve**

If the water temperature must not go beyond a maximum temperature  $\theta_{max}$  then the equation with the time as a variable is:

$$\theta_{max} = I^2 r \cdot R \left( 1 - e^{-\frac{t}{T_{th}}} \right) + \theta_A \quad (A11)$$

Solving for  $t$  gives the time-current equation:

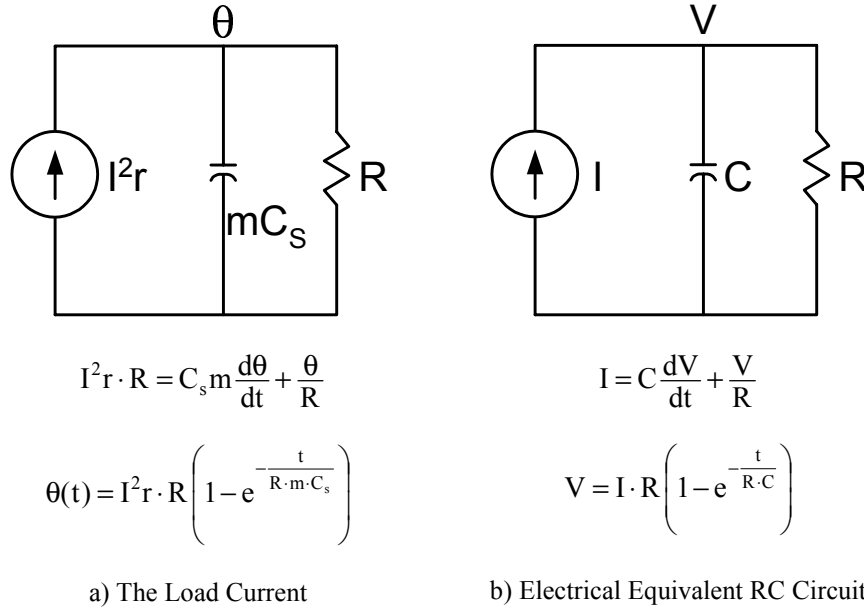
$$t = T_{th} \ln \left( \frac{I^2 r \cdot R}{I^2 r \cdot R - (\theta_{max} - \theta_A)} \right) \quad (A12)$$

Let us define the current  $I_{max}$  as the maximum current that can be supplied to the heating resistor without the water reaching maximum temperature as time goes to infinity. This maximum current would have to satisfy Equation A10 as in:

$$\theta_{max} = I_{max} r \cdot R + \theta_A \quad (A13)$$

or:

$$I_{max}^2 r \cdot R = \theta_{max} - \theta_A \quad (A14)$$



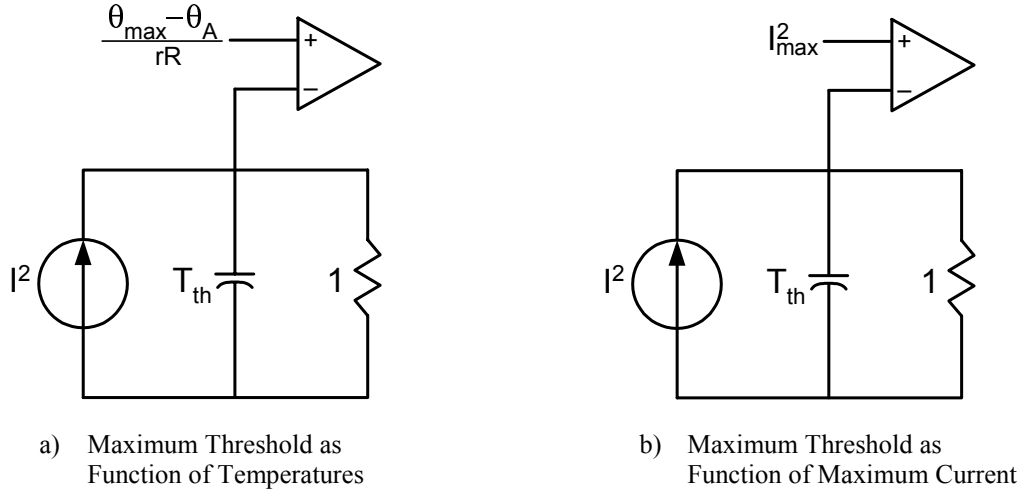
**Figure A2** Equivalence Between a Thermal System and a Parallel RC Circuit

Substituting  $\theta_{max} - \theta_A$  in Equation A12 gives the equation:

$$t = T_{th} \ln \left( \frac{I^2 r \cdot R}{I^2 r \cdot R - I_{max}^2 r \cdot R} \right) = T_{th} \ln \left( \frac{I^2}{I^2 - I_{max}^2} \right) \quad (A15)$$

In Equation A15, we have finally expressed the time to reach the hot-spot temperature as a function of the current. Equation A15 is also remarkable because we have removed all the temperature constants and replaced them with the maximum current  $I_{max}$ .

Based on Equations A11 and A15, the electrical equivalent circuit for the thermal process can be represented by the RC circuit as shown in Figure A2.



**Figure A3** Electrical RC Equivalent of Thermal Process

It should be noted that Equation A15 has no solution unless:

$$I > I_{\max} \quad (\text{A16})$$

Any current less than  $I_{\max}$  will raise the water temperature to a steady temperature given by Equation A7. This temperature will be represented by a capacitor voltage in the equivalent circuit of Figure A2(b).

In Equation A15, the time to maximum temperature is expressed implicitly with reference to the ambient temperature or with the initial current equal to zero. Let us assume we want to develop an expression for the time to hot-spot temperature when the steady-state current is the operating current  $I_{\text{op}}$ . In Equation A17, the time to maximum temperature starts with the temperature at ambient (or with the current supplied at zero value). With the newer equation, the time to maximum temperature starts with the temperature at operating or the current at the load current.

The time to reach the maximum temperature for some current  $I$  from the operating current (or temperature) is equal to the time to reach the maximum current from ambient with the same current minus the time to reach the operating temperature from ambient with the same current.

We can compute  $\theta_{\text{op}}$  from Equation A9:

$$\theta_{\text{op}} = I_{\text{op}}^2 r \cdot R + \theta_A \quad (\text{A17})$$

The time to reach the operating temperature  $t_{\text{op}}$  from ambient for a current  $I$  can be computed from Equation A8 as:

$$\theta_{\text{op}} = I^2 r \cdot R \left( 1 - e^{-\frac{t_{\text{op}}}{T_{\text{th}}}} \right) + \theta_A \quad (\text{A18})$$

from which we get  $t_{op}$  as:

$$t_{op} = T_{th} \ln \left( \frac{I^2 r \cdot R}{I^2 r \cdot R - (\theta_{op} - \theta_A)} \right) \quad (A19)$$

Replacing  $\theta_{op}$  by its value, we get:

$$t = T_{th} \ln \left( \frac{I^2 r \cdot R}{I^2 r \cdot R - I_{op}^2 r \cdot R} \right) = T_{th} \ln \left( \frac{I^2}{I^2 - I_{op}^2} \right) \quad (A20)$$

Finally, the time from operating current or temperature is provided by:

$$t = T_{th} \ln \left( \frac{I^2}{I^2 - I_{max}^2} \right) - T_{th} \ln \left( \frac{I^2}{I^2 - I_{op}^2} \right) = T_{th} \ln \left( \frac{I^2 - I_{op}^2}{I^2 - I_{max}^2} \right) \quad (A21)$$

This latter expression provides the time to reach the hot-spot temperature for a current  $I$  when starting from the operating current or operating temperature.

Normalizing the current with respect to the operating current, we get finally:

$$t = T_{th} \ln \left( \frac{I_{pu}^2 - 1}{I_{pu}^2 - SF^2} \right) \quad (A22)$$

Where SF, or the service factor, is defined as the ratio of the maximum current over the operating current:

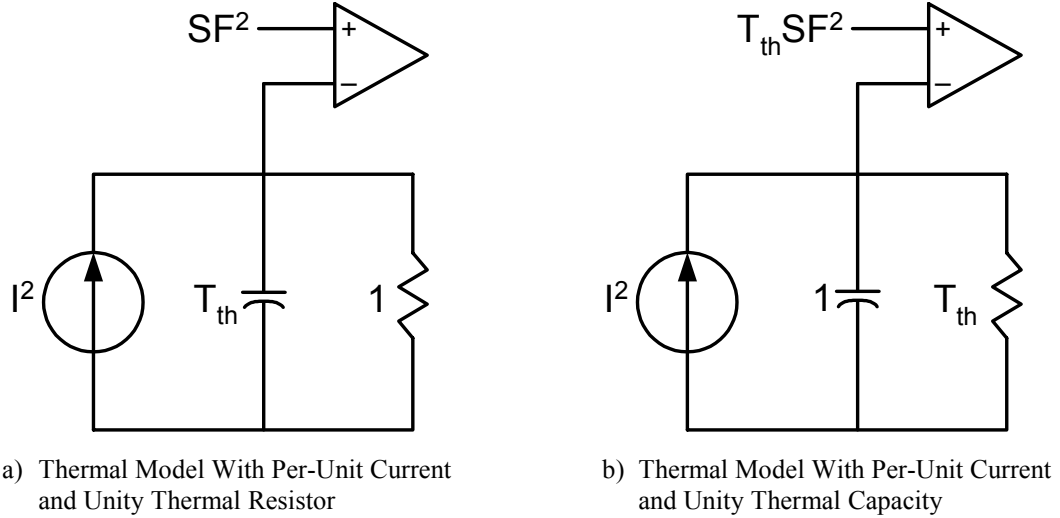
$$SF = \frac{I_{max}}{I_{op}} \quad (A23)$$

The electrical RC equivalent of the thermal process represented by Equation A22 is shown in Figure A4.

Note that the models in Figure A4 are identical to the models in Figure A3 except that the current is expressed in per unit. In Equation A21, time is measured starting from an initial operating temperature. This condition is represented in the model of Figure A4 by a voltage across the capacitor equivalent to the initial operating temperature.

Starting from Equation A11, we can define the following invariant quantity:

$$\text{Constant} = \theta - \theta_A = I^2 r \cdot R \left( 1 - e^{-\frac{t}{T_{th}}} \right) \quad (A24)$$



**Figure A4** Electrical RC Equivalent of Thermal Process

Equation A24 indicates that for any pair of points  $(I_0, T_0)$  and  $(I_1, T_1)$  on the time-current curve, we have the following identity:

$$\text{Constant} = I_1^2 r \cdot R \left( 1 - e^{-\frac{t_1}{T_{th}}} \right) = I_0^2 r \cdot R \left( 1 - e^{-\frac{t_0}{T_{th}}} \right) \quad (\text{A25})$$

Equation A25 can otherwise be expressed as:

$$\frac{\left( 1 - e^{-\frac{t_1}{T_{th}}} \right)}{\left( 1 - e^{-\frac{t_0}{T_{th}}} \right)} = \frac{I_0}{I_1} \quad (\text{A26})$$

Relations A25 and A26 determine if a given time-current curve corresponds to a first-order thermal model.

## REFERENCES

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- [2] IEEE Std 620-1996, *Guide for the Presentation of Thermal Limit Curves for Squirrel Cage Induction Motors*.
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## BIOGRAPHIES

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His biography appears in Who's Who in America. He holds over a dozen patents associated with power system protection using solid state and microprocessor technology and is the author of numerous IEEE and Protective Relay Conference papers. He received the Best Paper Award of the 1988 Petroleum and Chemical Industry Conference and the Power System Relaying Committee's Distinguished Service Award in 1991.

**Gabriel Benmouyal** received his B.A.Sc. in Electrical Engineering and his M.A.Sc. in Control Engineering from Ecole Polytechnique, Université de Montréal, Canada, in 1968 and 1970, respectively. In 1969, he joined Hydro-Québec as an instrumentation and control specialist. He worked on different projects in the field of substation control systems and dispatching centres. In 1978, he joined IREQ, where his main field of activity was the application of microprocessors and digital techniques to substation and generating-station control and protection systems. In 1997, he joined Schweitzer Engineering Laboratories in the position of Research Engineer. He is a registered professional engineer in the Province of Québec, is an IEEE member, and has served on the Power System Relaying Committee of IEEE since May 1989.